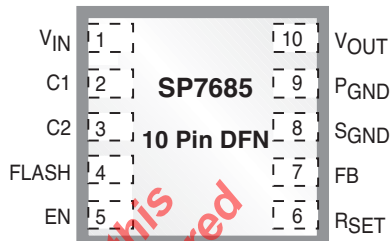


1.2A Buck/Boost Charge Pump LED Driver

FEATURES

- Output Current up to 1.2A
- Up to 94% Efficiency in Torch Mode
- Adjustable FLASH Mode Current
- Minimum External Components: No Inductors
- Automatic Buck/Boost Mode Switchover
- Wide V_{IN} Range: 2.7V to 5.5V
- High Frequency Operation: 2.4 MHz
- 50mV Reference for low Loss Sensing
- $I_Q < 2\mu A$ in Shutdown
- PWM Dimming Control
- Automatic Soft Start Limits Inrush Current
- Overvoltage Protection on Output
- Overcurrent/temperature Protection
- Low Ripple and EMI
- Ultra-low Dropout Voltage in Buck Mode
- 2.6 Second Timeout in Flash Mode
- Space Saving RoHS Compliant, Lead Free Package: 10-pin 3mm x 3mm DFN



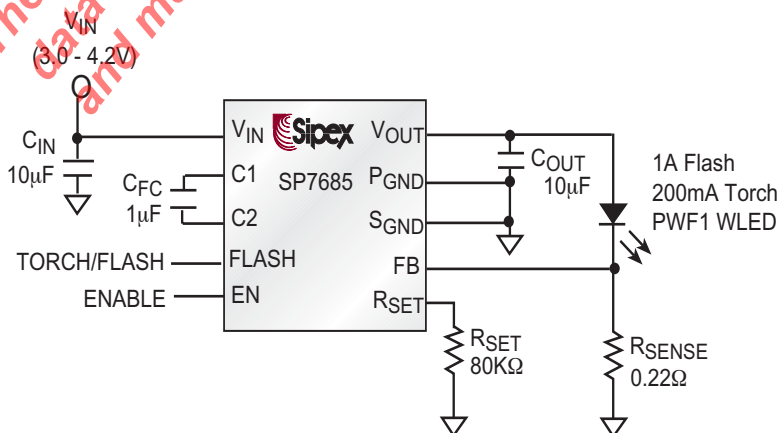
APPLICATIONS

- White LED Torch/Flash for Cell Phones, DSCs, and Camcorders
- White LED Backlighting
- Generic Lighting/Flash/Strobe Applications
- General Purpose High Current Boost

DESCRIPTION

The SP7685 is a current-regulated charge pump ideal for powering high brightness LEDs for camera flash applications. The charge pump can be set to regulate two current levels for FLASH and TORCH modes. The SP7685 automatically switches modes between step-up and step-down ensuring that LED current does not depend on the forward voltage. A low current sense reference voltage (50mV) allows the use of small 0603 current sensing resistors. The SP7685 is offered in 10-pin DFN package.

TYPICAL APPLICATION CIRCUIT



ABSOLUTE MAXIMUM RATINGS

V_{IN}, V_{OUT}	-0.3V to 6V
Output Current Pulse (Flash)	2A
Output Current Continuous (Torch)	0.4A
Storage Temperature	-65°C to +150°C
Operating Temperature	-40°C to +85°C
V_{EN}	0.0V to 7V
3x3 10 DFN	$\Theta_{JA} = 40.5^{\circ}\text{C/W}$
ESD Rating.	2kV HBM

These are stress ratings only and functional operation of the device at these ratings or any other above those indicated in the operation sections of the specifications below is not implied. Exposure to absolute maximum rating conditions for extended periods of time may affect reliability.

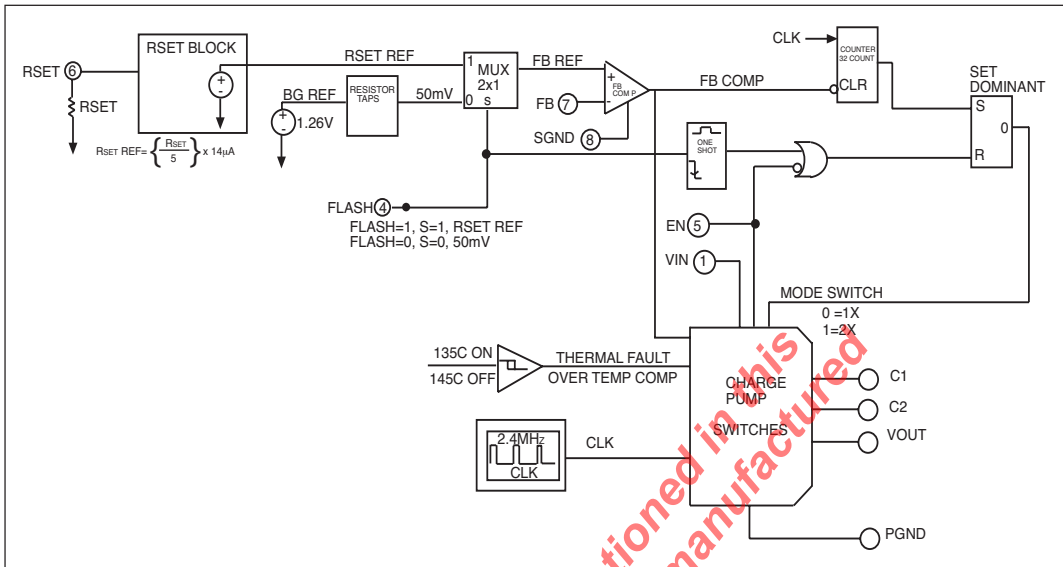
ELECTRICAL CHARACTERISTICS

$T_A = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$, $V_{IN} = 3.6\text{V}$, $C_{IN} = 10\mu\text{F}$, $C_{FC} = 1.0\mu\text{F}$, $C_{OUT} = 10\mu\text{F}$. $V_{SHDN} = V_{IN}$, typical values at 25°C . The ♦ denotes the specifications which apply over the full operating temperature range unless otherwise noted.

PARAMETER	MIN.	TYP.	MAX.	UNITS	CONDITIONS
Operating Input Voltage	2.7		5.5	V	♦
Quiescent Current		0.5	3	mA	♦ $V_{IN} = 2.7 - 5.5\text{V}$ FLASH = GND, 1X Mode, $I_{LOAD} = 100\mu\text{A}$
		2			FLASH = High, 2x mode
Shutdown Current			2	μA	$V_{IN} = 5.5\text{V}$, $V_{EN} = 0.0\text{V}$
Oscillator Frequency		2.4		MHz	
Charge Pump Equivalent Resistance (x2 mode)		4		Ω	$V_{FB} = 0.0\text{V}$, $V_{IN} = 3.6\text{V}$
Charge Pump Equivalent Resistance (x1 mode)		0.4	0.7	Ω	$V_{IN} = 3.6\text{V}$
FB Reference Voltage	45	50	55	mV	♦ FLASH = GND
	138	150	162	mV	♦ FLASH = High, $R_{SET} = 53.6\text{k}\Omega$
FB Reference Voltage Range	100		400	mV	♦ FLASH = High. Guaranteed by design.
FB Pin Current			0.5	μA	$V_{FB} = 0.3\text{V}$
EN, FLASH Logic Low			0.4	V	♦
EN, FLASH Logic High	1.3			V	♦
EN, FLASH Pin Current			0.5	μA	♦
V_{OUT} Turn-on Time		170	500	μs	♦ $V_{IN} = 3.6\text{V}$, FB within 90% of regulation
Thermal Shutdown Temperature		145		$^{\circ}\text{C}$	
Maximum Flash ON time	1.6	2.6	3.6	s	♦ FLASH = High

PIN NUMBER	PIN NAME	DESCRIPTION
1	V _{IN}	Input Voltage for the charge pump. Decouple with 4.7 μ F ceramic capacitor close to the pins of the IC.
2	C1	Positive input for the external flying capacitor. Connect a ceramic 1 μ F capacitor close to the pins of the IC.
3	C2	Negative input for the external flying capacitor. Connect a ceramic 1 μ F capacitor close to the pins of the IC.
4	FLASH	Logic input to toggle operation between FLASH and TORCH mode. In TORCH mode FB is regulated to the internal 50mV reference. In FLASH mode FB reference voltage can be adjusted by changing the resistor from R _{SET} pin to ground. Choose the external current sense resistor (R _{SENSE}) based on desired current in TORCH mode. This pin does not have an internal pull-up/pull-down; do not leave this pin floating.
5	EN	Shutdown control input. Connect to V _{IN} for normal operation, connect to ground for shutdown. This pin does not have an internal pull-up/pull-down; do not leave this pin floating.
6	R _{SET}	Connect a resistor from this pin to ground. When in FLASH mode (FLASH = High) this resistor sets the current regulation point according to the following: V _{FB} = R _{SET} *14 μ A/5 (Flash Mode)
7	FB	Feedback input for the current control loop. Connect directly to the current sense resistor.
8	S _{GND}	Internal ground pin. Control circuitry returns current to this pin.
9	P _{GND}	Power ground pin. Flying capacitor current returns through this pin.
10	V _{OUT}	Charge Pump Output Voltage. Decouple with an external capacitor. At least 1 μ F is recommended. Higher capacitor values reduce output ripple

FUNCTIONAL DIAGRAM



THEORY OF OPERATION

The SP7685 is a charge pump regulator designed for converting a Li-Ion battery voltage of 2.7V to 4.2V to drive a white LED used in digital still camera Flash and Torch applications. The SP7685 has two modes of operation which are pin-selectable for either Flash or Torch. Flash mode is usually used with a pulse of about 200 to 300 milliseconds to generate a high intensity Flash. Torch can be used continuously at a lower output current than Flash and is often used for several seconds in a digital still camera "movie" mode.

The SP7685 also has two modes of operation to control the output current: the 1X mode and 2X mode. Operation begins after the enable pin EN receives a logic high, the bandgap reference wakes up after 200μs, and then SP7685 goes through a soft-start mode designed to reduce inrush current. The SP7685 starts in the 1X mode, which

acts like a linear regulator to control the output current by continuously monitoring the feedback pin FB. In 1X mode, if the SP7685 auto detects a dropout condition, which is when the FB pin is below the regulation point for more than 32 cycles of the internal clock, the SP7685 automatically switches to the 2X mode. The SP7685 remains in the 2X mode until one of four things happens: 1) the enable pin EN has been toggled, 2) the Flash pin has changed from high to low, 3) V_{IN} is cycled or, 4) a thermal fault occurs.

The 2X mode is the charge pump mode where the output can be pumped as high as two times the input voltage, provided the output does not exceed the maximum voltage for the SP7685, which is internally limited to about 5.5V. In the 2X mode, as in the 1X mode, the output current is regulated by the voltage at the FB pin.

In the Torch mode, (Flash = GND) the Flash pin is set to logic low and the SP7685 FB pin regulates to 50mV output:

$$V_{FB} = 50\text{mV (Torch Mode)}$$

When in Flash mode, (Flash = V_{IN}), the FB regulation voltage is set by the resistor R_{SET} connected between the R_{SET} pin and S_{GND} and the equation:

$$V_{FB} = R_{SET} * 14\mu\text{A} / 5 \text{ (Flash Mode)}$$

Where $14\mu\text{A}$ is an internal regulated current and 5 is an internal factor used to scale the V_{SET} voltage to the V_{FB} voltage. Typical values of R_{SET} are $140\text{K}\Omega$ to $35\text{K}\Omega$ for a range of $V_{FB} = 400\text{mV}$ to 100mV in Flash mode.

The output current is then set in either Flash or Torch mode by the equation:

$$I_{OUT} = V_{FB} / R_{SENSE}$$

FLASH TIMEOUT PROTECTION

Due to the high currents typically available in Flash mode, it is necessary to protect the white LED from damage if left on too long. The SP7685 has a timeout in Flash mode of approximately 2.6 seconds after which it will shut down operation. Operation will not begin again in Flash mode until the Enable pin or Flash pin have been set Low and then High again.

OVERTEMPERATURE PROTECTION

When the temperature of the SP7685 rises above 145°C , the overtemperature protec-

tion circuitry turns off the output switches to prevent damage to the device. If the temperature drops back down below 135°C , the part automatically recovers and executes a soft start cycle.

OVERVOLTAGE PROTECTION

The SP7685 has over voltage protection. If the output voltage rises above the 5.5V threshold, the over voltage protection shuts off all of the output switches to prevent the output voltage from rising further. When the output decreases below 5.5V , the device resumes normal operation.

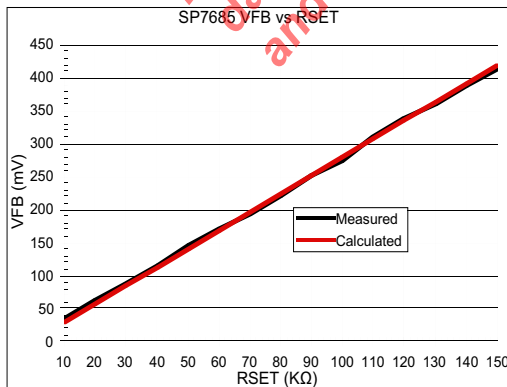
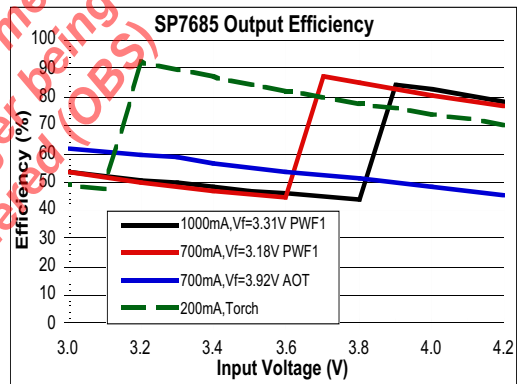
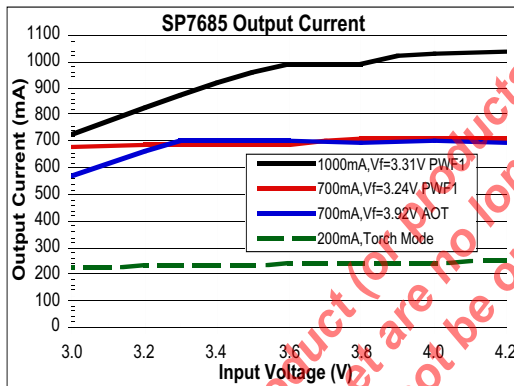
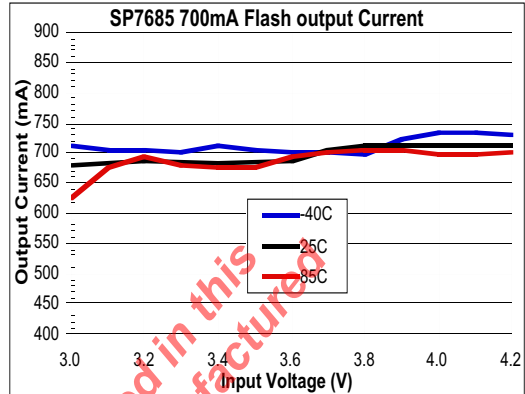
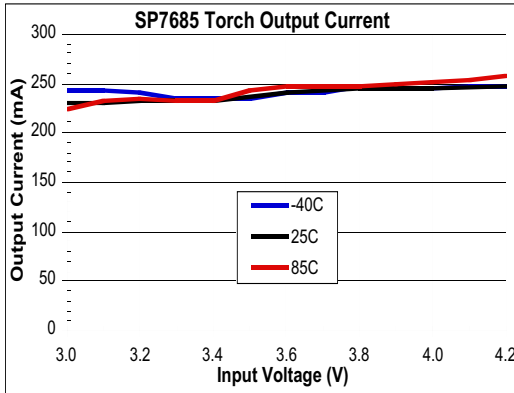
OVERCURRENT PROTECTION

The over current protection circuitry monitors the average current out of the V_{OUT} pin. If the average output current exceeds approximately 1.6 Amps, then the overcurrent protection circuitry shuts off the output switches to protect the chip.

BRIGHTNESS CONTROL USING PWM

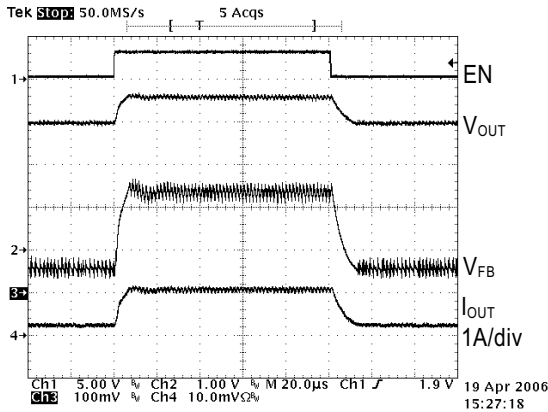
Dimming control can be achieved by applying a PWM control signal to the EN pin. The brightness of the white LEDs is controlled by increasing and decreasing the duty cycle of the PWM signal. While the operating frequency range of the PWM control is from 60Hz to 700Hz , the recommended maximum brightness frequency range of the PWM signal is from 60Hz to 200Hz . A repetition rate of at least 60Hz is required to prevent flicker.

$V_{IN}=3.6V$, Typical Application Circuit, $T_A = 25^\circ C$ unless otherwise noted.

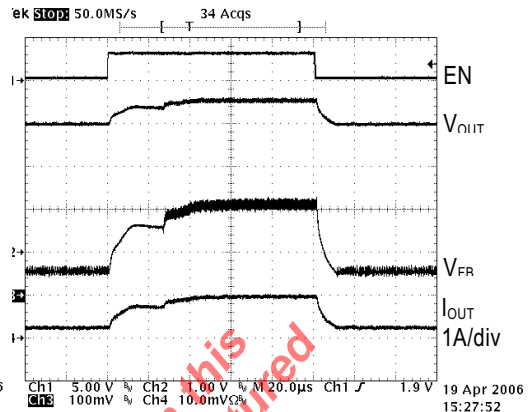


APPLICATION INFORMATION

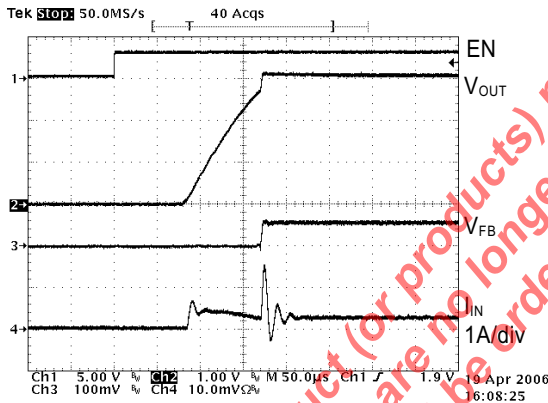
$V_{IN}=3.6V$, Typical Application Circuit, $T_A = 25^{\circ}C$ unless otherwise noted.



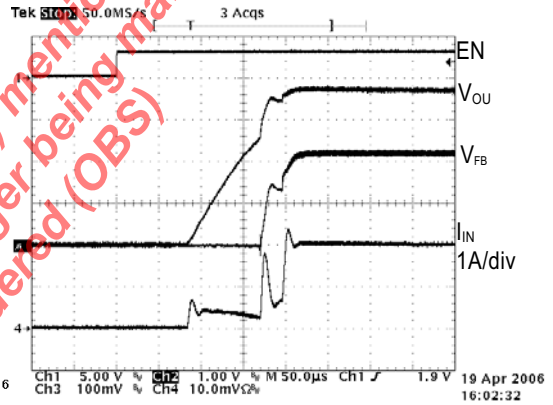
Torch in 1X to Flash in 1X Mode, $V_{IN}=4.2V$



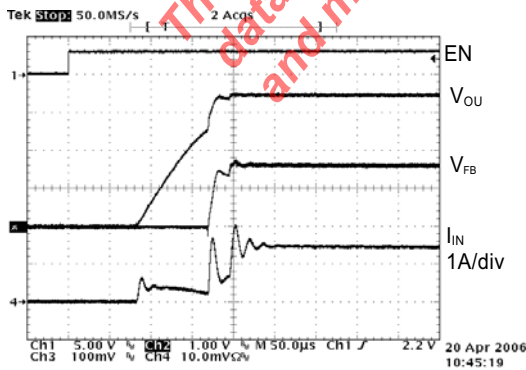
Torch in 1X to Flash in 2X Mode, $V_{IN}=3.6V$



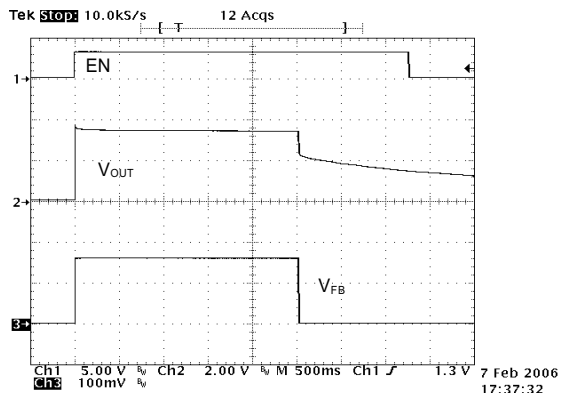
Start Up 200mA Torch
 $V_{IN}=3.6V$, $V_{OUT}=3.1V$



Start Up 1000mA Flash
 $V_{IN}=3.6V$, $V_{OUT}=3.8V$



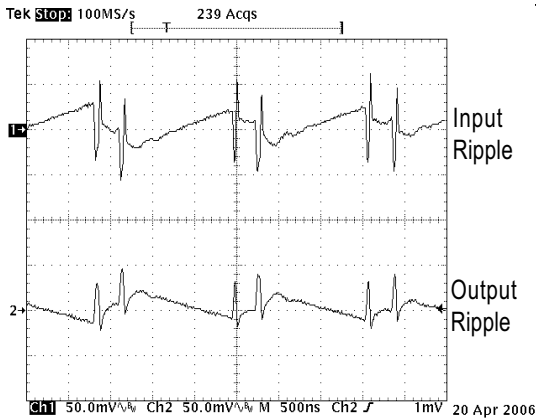
Start Up 700mA Flash
 $V_{IN}=3.6V$, $V_{OUT}=3.6V$



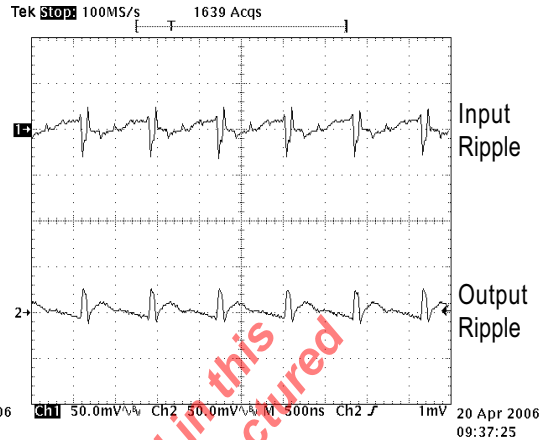
Flash Mode TimeOut Circuit at 2.6sec.
 $V_{IN}=4.2V$, $I_{OUT}=1A$

APPLICATION INFORMATION

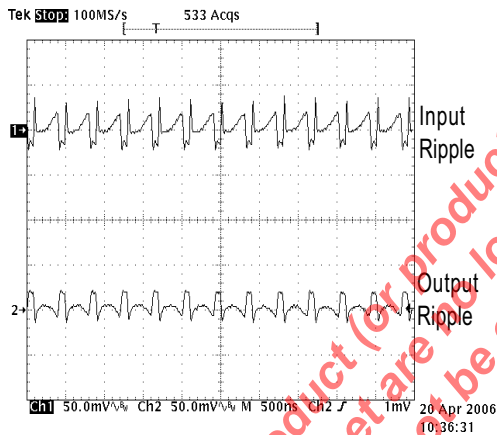
$V_{IN}=3.6V$, Typical Application Circuit, $T_A=25^{\circ}C$ unless otherwise noted.



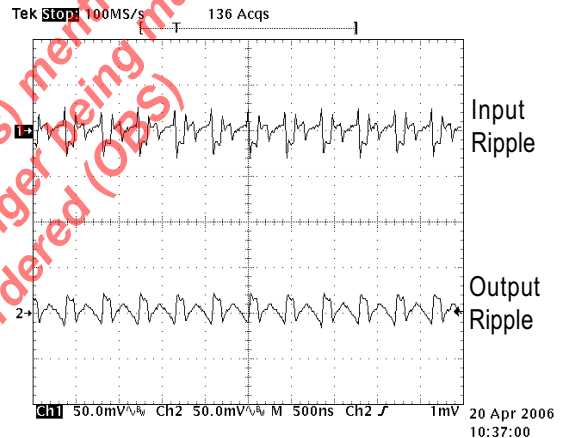
Ripple 1X Torch 200mA, $V_{IN}=4.2V$



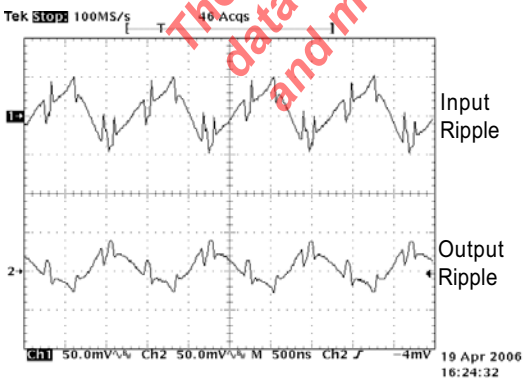
Ripple 2X Torch 200mA, $V_{IN}=3.0V$



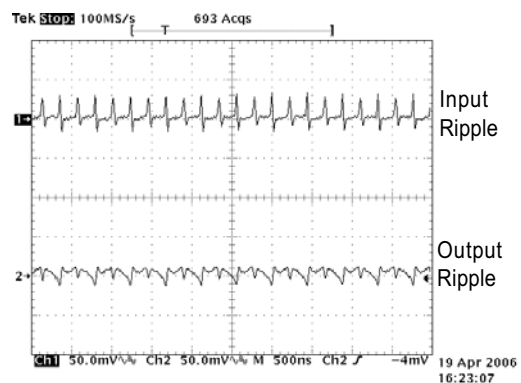
Ripple 1X Flash 700mA, $V_{IN}=4.2V$



Ripple 2X Flash 700mA, $V_{IN}=3.6V$



Ripple 1X Flash 1000mA, $V_{IN}=4.2V$



Ripple 2X Flash 1000mA, $V_{IN}=3.6V$

The SP7685 charge pump circuit requires three capacitors: 10 μ F input, 10 μ F output and 1 μ F fly capacitor are typically recommended. For the input capacitor, a value of 10 μ F will help reduce input voltage ripple for applications sensitive to ripple on the battery voltage. All the capacitors should be surface mount ceramic for low lead inductance necessary at the 2.4MHz switching frequency of the SP7685 and to obtain low ESR, which improves bypassing on the input and output and improves output voltage drive by reducing output resistance. Ceramic capacitors with X5R or X7R temperature grade are recommended for most applications. A selection of recommended capacitors is included in Table 1 below.

Manufacturer's Website	Part Number	Capacitance/ Voltage	Capacitor Size/ Type/Thickness	ESR @100kHz
TDK: www.tdk.com	C1005X5R0J105M	1 μ F/6.3V	0402/X5R/0.5mm	0.03
TDK: www.tdk.com	C1608X5R0J475K	4.7 μ F/6.3V	0603/X5R/0.9mm	0.02
TDK: www.tdk.com	C2012X5R0J106M	10 μ F/6.3V	0805/X5R/1.35mm	0.02
Murata: www.murata.com	GRM155R60J105KE19B	1 μ F/6.3V	0402/X5R/0.55mm	0.03
Murata: www.murata.com	GRM188R60J475KE19	4.7 μ F/6.3V	0603/X5R/0.9mm	0.02
Murata: www.murata.com	GRM21BR60J106KE19L	10 μ F/6.3V	0805/X5R/1.35mm	0.02

Table 1: Recommended Capacitors

The input and output capacitors should be located as close to the V_{IN} and V_{OUT} pins as possible to obtain best bypassing, and the returns should be connected directly to the P_{GND} pin or to the thermal pad ground located under the SP7685. The fly capacitor should be located as close to the C1 and C2 pins as possible. See typical circuit layout at the end of this section for details on the recommended layout.

To obtain low output ripple, a value of 10 μ F is recommended for C_{OUT} . For output currents of 500mA to 1.2A, the recommended C_{FC} fly capacitor value of 1 μ F should be used. Output currents in Flash of 100mA to 400mA can use a 0.47 μ F C_{FC} but a minimum 4.7 μ F C_{OUT} is still needed.

RESISTOR SELECTION

The sense resistor R_{SENSE} is determined by the value needed in the Torch mode for the desired output current by the equation:

$$R_{SENSE} = V_{FB} / I_{OUT} \text{ where } V_{FB} = 50mV \text{ (Torch Mode)}$$

Once the R_{SENSE} resistor has been selected for Torch mode, the V_{FB} voltage can be selected for Flash mode using the following equation:

$$V_{FB} = I_{OUT} * R_{SENSE} \text{ (Flash Mode) where } I_{OUT} \text{ is for Flash Mode.}$$

Next, the R_{SET} resistor can be selected for Flash mode using the following equation:

$$R_{SET} = \left(\frac{V_{FB}}{14\mu A} \right) * 5 \Omega \text{ (Flash Mode)}$$

For an example of 200mA Torch mode and 700mA Flash mode, the values $R_{SENSE} = 0.22\Omega$, $V_{FB} = 155mV$ (Flash Mode), and $R_{SET} = 56K\Omega$ are calculated. The power obtained in the Flash mode would be:

$$P_{FLASH} = V_{FB} * I_{OUT} = 155mV * 700mA = 109mW.$$

The typical 0603 surface mount resistor is rated 1/10 Watt continuous power and 1/5

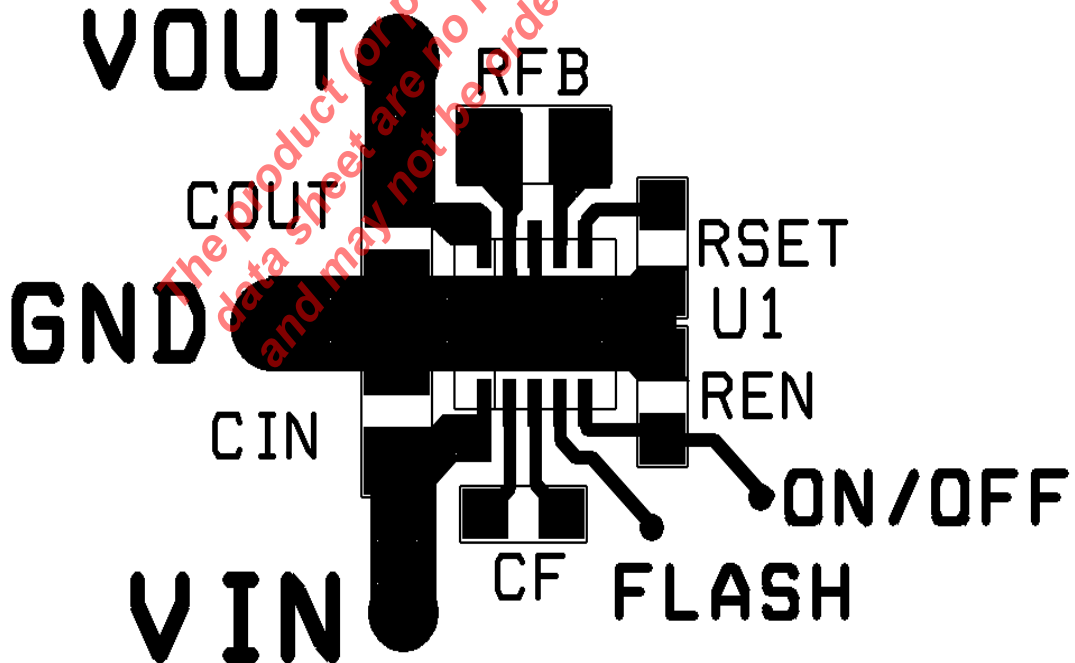
Watt pulsed power, more than enough for this application. For other applications, the P_{FLASH} power can be calculated and resistor size selected. The R_{SENSE} resistor is recom-

mended to be size 0603 for most applications. The range of typical resistor values and sizes are shown here in Table 2.

Part Reference	Value	Tolerance	Size	Manufacturers
RSET	33k Ω	5%	0402	any
RSET	39k Ω	5%	0402	any
RSET	43k Ω	5%	0402	any
RSET	47k Ω	5%	0402	any
RSET	56k Ω	5%	0402	any
RSET	62k Ω	5%	0402	any
RSET	68k Ω	5%	0402	any
RSET	82k Ω	5%	0402	any
RSET	100k Ω	5%	0402	any
RSET	110k Ω	5%	0402	any
RSET	120k Ω	5%	0402	any
RSET	150k Ω	5%	0402	any
RSENSE	0.22 Ω	5%	0603	Panasonic or Vishay
RSENSE	0.27 Ω	5%	0603	Panasonic or Vishay
RSENSE	0.33 Ω	5%	0603	Panasonic or Vishay
RSENSE	0.39 Ω	5%	0603	Panasonic or Vishay
RSENSE	0.47 Ω	5%	0603	Panasonic or Vishay
RSENSE	0.56 Ω	105%	0604	Panasonic or Vishay
RSENSE	0.68 Ω	205%	0605	Panasonic or Vishay

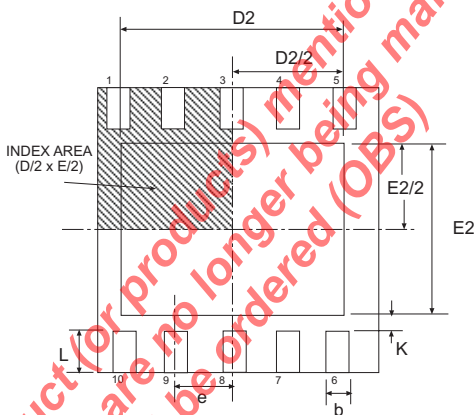
Table 2: Resistor values and sizes

EVALUATION BOARD CIRCUIT LAYOUT



Side view diagram of a chair. The diagram shows a cross-section of the chair with a backrest and a seat. The backrest is a trapezoid with a top horizontal edge, a bottom horizontal edge, and two slanted sides. The seat is represented by a horizontal line with four rectangular blocks underneath it. The following dimensions are labeled: A is the height from the seating plane to the top of the backrest; A_1 is the height from the seating plane to the top of the backrest on the right side; A_3 is the height from the seating plane to the bottom of the backrest; and θ is the angle between the slanted side of the backrest and the vertical line. The text "SEATING PLANE" is written below the seat, and "SIDE VIEW" is written below the diagram.

SIDE VIEW



BOTTOM VIEW

data sheet

3x3 10 Pin DFN			JEDEC MO-229			VARIATION VEED-5		
SYMBOL	Dimensions in Millimeters: Controlling Dimension			Dimensions in Inches Conversion Factor: 1 Inch = 25.40 mm				
	MIN	NOM	MAX	MIN	NOM	MAX		
A	0.80	0.90	1.00	0.032	0.036	0.039		
A1	0.00	0.02	0.05	0.000	0.001	0.002		
A3	0.20 REF			0.008 REF				
K	0.20	-	-	0.008	-	-		
ø	0°	-	14°	0°	-	14°		
b	0.18	0.25	0.30	0.008	0.010	0.012		
D	3.00 BSC			0.119 BSC				
D2	2.20	-	2.70	0.087	-	0.106		
E	3.00 BSC			0.119 BSC				
E2	1.40	-	1.75	0.056	-	0.069		
e	0.50 BSC			0.020 BSC				
L	0.30	0.40	0.50	0.012	0.016	0.020		
SIPEX Pkg Signoff Date/Rev:				JL Aug09-05 / RevA				

Part Number	Operating Temperature Range	Package Type
SP7685ER-L	-40°C to +85°C	(Lead Free) 10 Pin DFN
SP7685ER-L/TR	-40°C to +85°C	(Lead Free) 10 Pin DFN

/TR = Tape and Reel

Pack quantity is 3,000 for DFN.

*The product (or products) mentioned in this
data sheet are no longer being manufactured
and may not be ordered (OBS)*

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