Data Sheet

CDK3404 8-bit, 180MSPS, Triple Video DAC



FEATURES

- 8-bit resolution, 180MSPS
- ±2.5% gain matching
- ±0.5% linearity error
- Sync and blank controls
- $1.0V_{pp}$ video into 37.5Ω or 75Ω load
- Internal bandgap voltage referender
- Low glitch energy
- Single +3.3V power supply

APPLICATIONS

- Video signal conversion
 - RGB
 - YC_BC_R
- Composite, Y,
- Multimedia systems
- Image processing
- PC Graphics

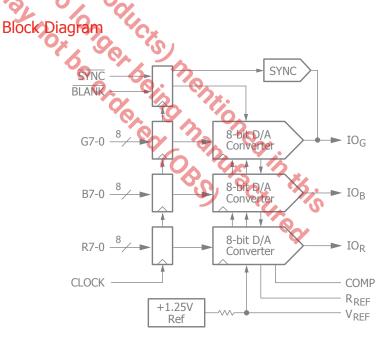
General Description

CDK3404 is a low-cost triple D/A converters that are tailored to fit graphics and video applications where speed is critical.

CMOS-level inputs are converted to analog current outputs that can drive 25-37.5 Ω loads corresponding to doubly-terminated 50-75 Ω loads. A sync current following $\overline{\text{SYNC}}$ input timing is added to the IO_G output. $\overline{\text{BLANK}}$ will override RGB inputs, setting IO_G, IO_B and IO_R currents to zero when $\overline{\text{BLANK}}$ = L. Although appropriate for many applications, the internal 1.25V reference voltage can be overridden by the V_{REF} input.

Few external components are required, just the current reference resistor, current output load resistors, bypass capacitors, and decoupling capacitors.

Package is a 48-lead TQFP. Fabrication technology is CMOS. Performance is guaranteed from 40° C to $+125^{\circ}$ C (CDK3404A) and 0° C to 70° C (CDK3404C).



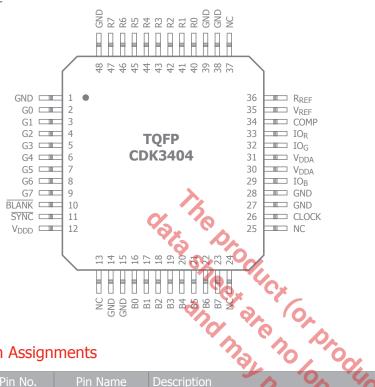
Ordering Information

| Part Number | Package | Pb-Free | RoHS Compliant | Operating Temp Range | Packaging Method | Package Quantity |
|--------------|---------|---------|----------------|----------------------|------------------|------------------|
| CDK3404CTQ48 | TQFP-48 | Yes | Yes | 0°C to +70°C | Tray | 250 |
| CDK3404ATQ48 | TQFP-48 | Yes | Yes | -40°C to +125°C | Tray | 250 |

Moisture sensitivity level for all parts is MSL-3.

Pin Configuration





Pin Assignments

| i in Abolgh | | |
|-------------------------------------|------------------|--------------------------------|
| Pin No. | Pin Name | Description |
| Clock and P | Pixel I/O | |
| 26 | CLK | Clock Input |
| 47-40 | R7-0 | Red Pixel Data Inputs |
| 9–2 | G7-0 | Green Pixel Data Inputs |
| 23–16 | B7-0 | Blue Pixel Data Inputs |
| Controls | | a ap a. |
| 11 | SYNC | Sync Pulse Input |
| 10 | BLANK | Blanking Input |
| Video Outp | uts | |
| 33 | IOR | Red Current Output |
| 32 | IOG | Green Current Output |
| 29 | IOB | Blue Current Output |
| Voltage Ref | ference | |
| 35 | V _{REF} | Voltage Reference Output/Input |
| 36 | R _{REF} | Current-Setting Resistor |
| 34 | COMP | Compensation Capacitor |
| Power and | Ground | |
| 30, 31 | V _{DDA} | Analog Power Supply |
| 12 | V _{DDD} | Digital Power Supply |
| 1, 14, 15, 27, 28, 38, 39, 48 | GND | Ground |
| 13, 24, 25, 37 | NC | No Connect |

Absolute Maximum Ratings

The safety of the device is not guaranteed when it is operated above the "Absolute Maximum Ratings". The device should not be operated at these "absolute" limits. Adhere to the "Recommended Operating Conditions" for proper device function. The information contained in the Electrical Characteristics tables and Typical Performance plots reflect the operating conditions noted on the tables and plots.

| Parameter | Min | Max | Unit |
|---|-------|------------------------|------|
| Power Supply Voltage | | | |
| V _{DDA} (Measured to GND) | -0.5 | 4.0 | V |
| V _{DDD} (Measured to GND) | -0.5 | 4.0 | V |
| Digital Inputs | | | |
| Applied Voltage (measured to GND) ⁽²⁾ | -0.5 | $V_{DDD} + 0.5$ | V |
| Forced Current ^(3,4) | -5.0 | 5.0 | mA |
| Analog Inputs | | | |
| Applied Voltage (measured to GND) ⁽²⁾ | -0.5 | V _{DDA} + 0.5 | V |
| Forced Current ^(3,4) | -10.0 | 10.0 | mA |
| Analog Outputs | | | |
| Applied Voltage (measured to GND) ⁽²⁾ | -0.5 | V _{DD} + 0.5 | V |
| Forced Current ^(3,4) | -60.0 | 60.0 | mA |
| Short Circuit Duration (single output in HIGH state to GND) | | unlimited | sec |

Reliability Information

| Parameter | Min | Max | Unit |
|--|------------|-----|------|
| Temperature | | | |
| Operating, Ambient | -40 | 125 | °C |
| Junction | | 150 | °C |
| Lead Soldering (10 seconds) | | 300 | °C |
| Vapor Phase Soldering (1 minute) | | 220 | °C |
| Storage | 6 5 | 150 | °C |
| Package Thermal Resistance (θ_{JA}) | 6.74 | 65 | °C/W |

Notes:

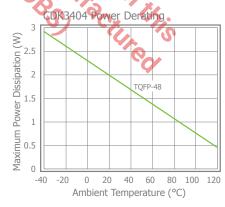
1. Functional operation under any of these conditions is NOT implied.

Performance and reliability are guaranteed only if Operating Conditions are not exceeded.

2. Applied voltage must be current limited to specified range.

3. Forcing voltage must be limited to specified range.

4. Current is specified as conventional current flowing into the device.



Recommended Operating Conditions

| Symbol | Parameter | Min | Тур | Max | Unit |
|------------------|---|-----|------|------|------|
| V _{DD} | Power Supply Voltage | 3.0 | 3.3 | 3.6 | V |
| V _{REF} | Reference Voltage, External | 1.0 | 1.25 | 1.5 | V |
| C _C | Compensation Capacitor | | 0.1 | | μF |
| R _L | Output Load | | 37.5 | | Ω |
| - | Ambient Temperature, Still Air (CDK3404A) | -40 | | +125 | °C |
| T _A | Ambient Temperature, Still Air (CDK3404C) | 0 | | +70 | °C |

Electrical Characteristics

 $(T_A = 25^{\circ}C, V_{DDA} = V_{DDD} = 3.3V, V_{REF} = 1.25V, R_L = 37.5\Omega$, unless otherwise noted)

| | | | Тур | Max | Units |
|--|--|---|---|--|--|
| | $T_A = 25^{\circ}C^{(1)}$ | | 80 | 85 | mA |
| Power Supply Current | $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(2)(3)}$ | | | 95 | mA |
| Total Power Dissipation ⁽²⁾ | $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ | | | 300 | mW |
| | | | | | |
| Input Voltage, HIGH (1) | $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ | 2.5 | | | V |
| Input Voltage, LOW (1) | $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ | | | 0.8 | V |
| Input Current, HIGH (1) | $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ | -1 | | 1 | μΑ |
| Input Current, LOW (1) | $T_{A} = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ | -1 | | 1 | μΑ |
| Input Capacitance | | | 4 | | pF |
| is 🔪 | | | | | |
| Output Current (1) | | | | 30 | mA |
| Output Resistance | | | 40 | | kΩ |
| Output Capacitance | | | 7 | | pF |
| put 😪 | <i>d</i> , | | | | |
| Reference Voltage Output (1) | $T_{\rm A} = -40^{\circ}{\rm C} \text{ to } +125^{\circ}{\rm C}^{(3)}$ | 1.135 | 1.25 | 1.365 | V |
| | Input Voltage, HIGH ⁽¹⁾ Input Voltage, LOW ⁽¹⁾ Input Current, HIGH ⁽¹⁾ Input Current, LOW ⁽¹⁾ Input Capacitance s Output Current ⁽¹⁾ Output Resistance Output Capacitance | Total Power Dissipation ⁽²⁾ $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ Input Voltage, HIGH ⁽¹⁾ $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ Input Voltage, LOW ⁽¹⁾ $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ Input Current, HIGH ⁽¹⁾ $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ Input Current, LOW ⁽¹⁾ $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ Input Capacitance Total Power Dissipation (1) S Output Current ⁽¹⁾ Output Capacitance Output Capacitance Put Total Power Dissipation (1) Reference Voltage Output ⁽¹⁾ Total Power Dissipation (1) | Total Power Dissipation ⁽²⁾ $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ Input Voltage, HIGH ⁽¹⁾ $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ 2.5 Input Voltage, LOW ⁽¹⁾ $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ -1 Input Current, HIGH ⁽¹⁾ $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ -1 Input Current, LOW ⁽¹⁾ $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ -1 Input Capacitance Output Current ⁽¹⁾ Output Current ⁽¹⁾ -1 Output Current ⁽¹⁾ Output Capacitance Output Capacitance -1 Put Total Power Capacitance Total Power Capacitance -1 Put Total Power Capacitance Total Power Capacitance -1 Instruct Capacitance Total Power Capacitance -1 -1 Number Capacitance Instruct Capacitance -1 -1 Instruct Capacitance -1 | Total Power Dissipation ⁽²⁾ $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ Image: Constraint of the system of the | Total Power Dissipation ⁽²⁾ $T_A = -40^{\circ}C \text{ to} + 125^{\circ}C^{(3)}$ Image: Second system 300 Input Voltage, HIGH ⁽¹⁾ $T_A = -40^{\circ}C \text{ to} + 125^{\circ}C^{(3)}$ 2.5 Input Voltage, LOW ⁽¹⁾ $T_A = -40^{\circ}C \text{ to} + 125^{\circ}C^{(3)}$ 2.5 0.8 Input Current, HIGH ⁽¹⁾ $T_A = -40^{\circ}C \text{ to} + 125^{\circ}C^{(3)}$ -1 1 1 Input Current, LOW ⁽¹⁾ $T_A = -40^{\circ}C \text{ to} + 125^{\circ}C^{(3)}$ -1 1 1 Input Capacitance 4 300 Output Current ⁽¹⁾ 30 30 30 Output Current ⁽¹⁾ 30 30 Output Current ⁽¹⁾ 30 30 Output Capacitance 30 30 Output Capacitance 30 30 Output Capacitance 30 30 Output Capacitance 30 30 30 30 3 |

Notes:

1. 100% tested at 25°C.

2. Parameter is guaranteed (but not tested) by design and characterization data

3. $T_A = -40^{\circ}C$ to $+125^{\circ}C$ for CDK3404A; $T_A = 0^{\circ}C$ to $+70^{\circ}C$ for CDK3404C.

Switching Characteristics

 $(T_A = 25^{\circ}C, V_{DDA} = V_{DDD} = 3.3V, V_{REF} = 1.25V, R_L - 37.5\Omega$, Griess otherwise noted)

| Symbol | Parameter | Conditions | Min | Тур | Max | Units |
|-------------------|--|---|------------|-----|-----|-------|
| Clock Input | | Or Cip Ch | | | | |
| | Conversion Rate (1) | $T_{A} = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ | | | 180 | MSPS |
| t _{PWH} | Pulse-width HIGH (2) | $T_{A} = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ | 2 | | | ns |
| t _{PWL} | Pulse-width LOW (2) | $T_A = -40^{\circ}C$ to $+125^{\circ}C$ | 2 | | | ns |
| Data Inputs | | | 15 12 | | | |
| | | $T_{A} = 25^{\circ}C^{(1)}$ | 1.5 | 5. | | ns |
| t _S | Setup | $T_{A} = -40^{\circ}C \text{ to } +125^{\circ}C^{(2)(3)}$ | 2 | 5 | | ns |
| | | T _A = 25°C ⁽¹⁾ | | | 0.6 | ns |
| t _H | Hold | $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(2)(3)}$ | ` Q | | 0.6 | ns |
| Data Output | s, with 50 Ω doubly terminated load | | | | | |
| t _D | Clock to Output Delay | $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ | | 1.6 | | ns |
| t _R | Output Risetime | $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ | | 0.6 | | ns |
| t _F | Output Falltime | $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(3)}$ | | 0.4 | | ns |
| t _{SET} | Settling Time | | | 2.5 | | ns |
| t _{SKEW} | Output Skew | | | 0.3 | | ns |

Notes:

1. 100% production tested at +25°C.

2. Parameter is guaranteed (but not tested) by design and characterization data.

3. T_A = -40°C to +125°C for CDK3404A; T_A = 0°C to +70°C for CDK3404C.

DC Performance

 $(T_A = 25^{\circ}C, V_{DDA} = V_{DDD} = 3.3V, V_{REF} = 1.25V, R_L = 37.5\Omega$, unless otherwise noted)

| Symbol | Parameter | Conditions | Min | Тур | Мах | Units |
|--------|------------------------------|---|-------|------|------|-------|
| | Resolution | | 8 | | | bits |
| TAU | | $T_{A} = 25^{\circ}C^{(1)}$ | -0.5 | | 0.5 | LSB |
| INL | Integral Linearity Error | $T_{A} = -40^{\circ}C \text{ to } +125^{\circ}C^{(2)(3)}$ | -0.5 | | 0.5 | LSB |
| DNI | | $T_{A} = 25^{\circ}C^{(1)}$ | -0.5 | | 0.5 | LSB |
| DNL | Differential Linearity Error | $T_{A} = -40^{\circ}C \text{ to } +125^{\circ}C^{(2)(3)}$ | -0.5 | | 0.5 | LSB |
| | Offset Error | $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(2)(3)}$ | | | 0.01 | %FS |
| | Gain Matching Error | $T_{A} = -40^{\circ}C \text{ to } +125^{\circ}C^{(1)}$ | -2.5 | | 2.5 | %FS |
| | Absolute Gain Error | $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(1)}$ | -3.5 | | 3.5 | %FS |
| | | $T_{A} = 25^{\circ}C^{(1)}$ | 18.0 | 18.7 | 19.4 | mA |
| | > | $T_A = -40^{\circ}C \text{ to } +125^{\circ}C^{(2)(3)}$ | 18.0 | 18.7 | 19.4 | mA |
| | Full-Scale Output Current | $T_A = -40$ °C to $+125$ °C , With internal reference. Trim RSET to calibrate full-scale current. | | 18.7 | | mA |
| PSRR | Power Supply Rejection Ratio | $T_{\rm A} = -40^{\circ}{\rm C} \text{ to } +125^{\circ}{\rm C}^{(2)(3)}$ | -0.01 | 0 | 0.01 | %/% |

Notes:

1. 100% production tested at +25°C.

2. Parameter is guaranteed (but not tested) by design and c

3. $T_A = -40^{\circ}$ C to +125°C for CDK3404A; $T_A = 0^{\circ}$ C to +7

AC Performance

 $(T_A = 25^{\circ}C, V_{DDA} = V_{DDD} = 3.3V, V_{REF} = 1.25V_{RL}$ Ω , unless otherwise noted)

| Symbol | Parameter | Condition | Min | Тур | Max | Units |
|-------------|--|---------------------|----------|-----|-----|-------|
| Analog Outp | puts | | l | | | |
| | Glitch Energy | | | 20 | | pVsec |
| | DAC-to-DAC Crosstalk | a nonti | | 30 | | dB |
| | Data Feedthrough | | | 50 | | dB |
| | Clock Feedthrough | | 0 | 60 | | dB |
| | uction tested at +25°C. s guaranteed (but not tested) by design and cha | racterization data. | Taciured | | | |

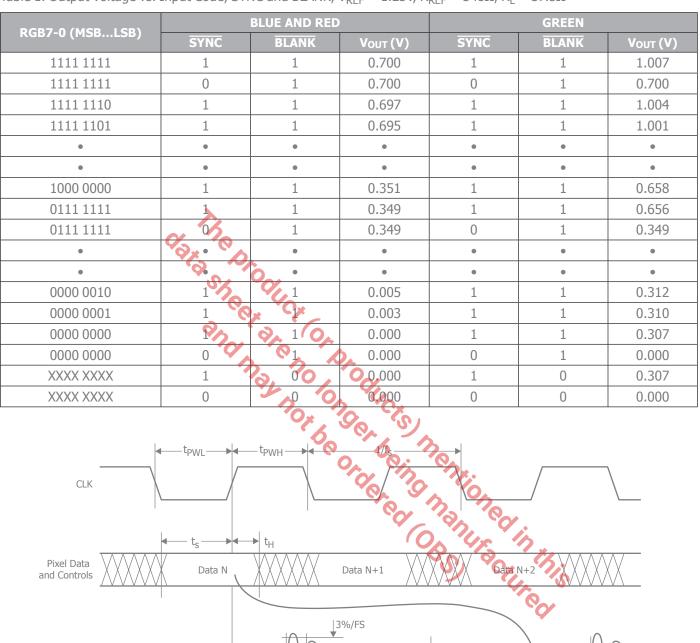


Table 1. Output Voltage vs. Input Code, \overline{SYNC} and \overline{BLANK} , $V_{REF} = 1.25V$, $R_{REF} = 348\Omega$, $R_L = 37.5\Omega$



Figure 1. CDK3404 Timing Diagram

tSET

tn

50%

90%

10%

t_R →

- t_F

Functional Description

Within the CDK3404 are three identical 8-bit D/A converters, each with a current source output. External loads are required to convert the current to voltage outputs. Data inputs RGB7-0 are overridden by the BLANK input. SYNC = H activates, sync current from I_{OS} for sync-on-green video signals.

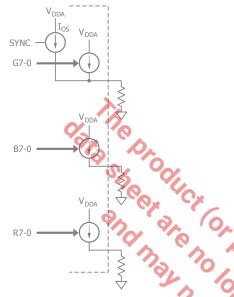


Figure 2. CDK3404 Current Source Structur

Digital Inputs

Incoming GBR data is regsitered on the rising edge of the clock input, CLK. Analog outputs follow the rising edge of CLK after a delay, t_{DO}.

Clock Input - CLK

Pixel data is registered on the rising edge of CLK. CLK should be driven by a dedicated buffer to avoid reflection induced jitter, overshoot, and undershoot.

Pixel Data Inputs - R7-0, B7-0, G7-0

RGB digital inputs are registered on the rising edge of CLK.

SYNC and BLANK

SYNC and BLANK inputs control the output level (Figure 3 and Table 1, on the previous page) of the D/A converters during CRT retrace intervals. BLANK forces the D/A outputs to the blanking level while SYNC = L turns off a current source, I_{OS}, that is connected to the green D/A converter. $\overline{\text{SYNC}}$ = H adds a 112/256 fraction of full-scale current to the green output. $\overline{SYNC} = L$ extinguishes the sync current during the sync tip.

 $\overline{\text{BLANK}}$ gates the D/A inputs. If $\overline{\text{BLANK}}$ = HIGH, the D/A inputs control the output currents to be added to the output blanking level. If $\overline{BLANK} = Low$, data inputs and the pedestal are disabled.

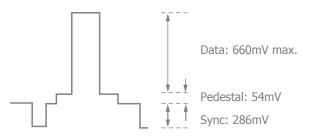


Figure 3. Normal Output Levels

Sync Pulse Input - SYNC

Blanking Input - BLANK When BLANK is LOW, MA converter outr raistered c Bringing SYNC LOW, disables a current source which superimposes a sync pulse on the IO_G output. SYNC and pixel data are registered on the rising edge of CLK. SYNC does not override any other data and should be used only during the blanking interval. If sync pulses are not re-Quired, SYNC should be connected to GND.

When **BLANK** is LOW, pixel data inputs are ignored and the DA converter outputs are driven to the blanking level. BLANK is registered on the rising edge of CLK.

D/A Outputs

Each D/A output is a current source from the V_{DDA} supply. Expressed in current units, the GBR transformation from data to current is as follows:

 $G = G7-0 \& \overline{BLANK} + \overline{SYNC} * 112$ $B = B7-0 \& \overline{BLANK}$

R = R7-0 & BLANK

Typical LSB current step is 73.2µA. To obtain a voltage output, a resistor must be connected to ground. Output voltage depends upon this external resistor, the reference voltage, and the value of the gain-setting resistor connected between R_{RFF} and GND.

To implement a doubly-terminated 75Ω transmission line, a shunt 75 Ω resistor should be placed adjacent to the analog output pin. With a terminated 75Ω line connected to the analog output, the load on the CDK3404 current source is 37.5Ω .

The CDK3404 may also be operated with a single 75Ω terminating resistor. To lower the output voltage swing to the desired range, the nominal value of the resistor on R_{REF} should be doubled.

R, G, and B Current Outputs - $\mathrm{IO}_{R^{\prime}}$ IO_{G} IO_{B}

Current source outputs can drive VESA VSIS, and RS-343A/SMPTE-170M compatible levels into doubly-terminated 75 Ω lines. Sync pulses can be added to the green output. When $\overline{\text{SYNC}}$ is HIGH, the current added to IO_G is:

 $IO_S = 2.33 (V_{REF}/R_{REF})$

Current-Setting Resistor - R_{REF}

Full-scale output current of each D/A converter is determined by the value of the resistor connected between R_{REF} and GND. Nominal value of R_{REF} is found from:

 $R_{REF} = 5.31 \; (V_{REF}/I_{FS})$

where I_{FS} is the full-scale (white) output current (in amps from the D/A converter (without sync). Sync 5 0.439 I_{FS} .

D/A full-scale (white) current may also be calculated fr

 $\mathrm{I}_{FS}=V_{FS}/R_L$

Where V_{FS} is the white voltage level and R_L is the total resistive load (Ω) on each D/A converter. V_{FS} is the blank to full-scale voltage.

Voltage Reference

Full scale current is a multiple of the current I_{SET} through an external resistor, R_{SET} connected between the R_{REF} pin

and GND. Voltage across R_{SET} is the reference voltage, $V_{REF,}$ which can be derived from either the 1.25 volt internal bandgap reference or an external voltage reference connected to V_{REF} . To minimize noise, a $0.1\mu F$ capacitor should be connected between V_{REF} and ground. I_{SET} is mirrored to each of the GBR output current sources. To minimize noise, a $0.1\mu F$ capacitor should be connected between the COMP pin and the analog supply voltage $V_{DDA}.$

Voltage Reference Output/Input - V_{REF}

An internal voltage source of +1.25V is output on the V_{REF} pin. An external +1.25V reference may be applied to override the internal reference. Decoupling V_{REF} to GND with a 0.1µF ceramic capacitor is required.

Power and Ground

Required power is a single +3.3V supply. To minimize power supply induced noise, analog +3.3V should be connected to V_{DDD} and V_{DDA} pins with 0.1μ F and 0.01μ F decoupling capacitors placed adjacent to each V_{DD} pin or pin pair.

The high slew-rate of digital data makes capacitive coupling to the outputs of any D/A converter a potential problem. Since the digital signals contain high-frequency components of the CLK signal, as well as the video output signal, the resulting data feedthrough often looks like harmonic distortion or reduced signal-to-noise performance. All ground pins should be connected to a common solid ground plane for best performance.

Applications Dicussion

Figure 5 (on the following page) illustrates a typical CDK3404 interface circuit. In this example, an optional 1.2V band-gap reference is connected to the V_{REF} output, overriding the internal voltage reference source.

Grounding

It is important that the CDK3404 power supply is wellregulated and free of high-frequency noise. Careful power supply decoupling will ensure the highest quality video signals at the output of the circuit. The CDK3404 has separate analog and digital circuits. To keep digital system noise from the D/A converter, it is recommended that power supply voltages come from the system analog power source and all ground connections (GND) be made to the analog ground plane. Power supply pins should be individually decoupled at the pin.

Printed Circuit Board Layout

Designing with high-performance mixed signal circuits demands printed circuits with ground planes. Overall system performance is strongly influenced by the board layout. Capacitive coupling from digital to analog circuits may result in poor D/A conversion. Consider the following suggestions when doing the layout:

- 1. Keep the critical analog traces (V_{REF} , I_{REF} , COMP, IO_S , IO_R , IO_G) as short as possible and as far as possible from all digital signals. The CDK3404 should be located near the board edge, close to the analog out-put connectors.
- 2. The power plane for the CDK3404 should be separate from that which supplies the digital circuitry. A single power plane should be used for all of the V_{DD} pins. If the power supply for the CDK3404 is the same as that of the system's digital circuitry, power to the CDK3404 should be decoupled with 0.1μ F and 0.01μ F capacitors and isolated with a ferrite bead.
- 3. The ground plane should be solid, not cross-hatched. Connections to the ground plane should have very short leads.

- 4. If the digital power supply has a dedicated power plane layer, it should not be placed under the CDK3404, the voltage reference, or the analog outputs. Capacitive coupling of digital power supply noise from this layer to the CDK3404 and its related analog circuitry can have an adverse effect on performance.
- CLK should be handled carefully. Jitter and noise on this clock will degrade performance. Terminate the clock line carefully to eliminate overshoot and ringing.

Improved Transisiton Times

Output shunt capacitance dominates slowing of output transition times, whereas series inductance causes a small amount of ringing that affects overshoot and settling time. With a doubly terminated 75 Ω load, transition times can be improved by matching the capacitive impedance output of the CDK3404. Output capacitance can be matched with a 220nH inductor in series with the 75 Ω source termination.

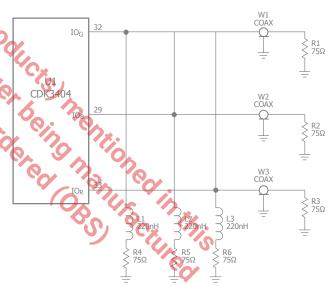
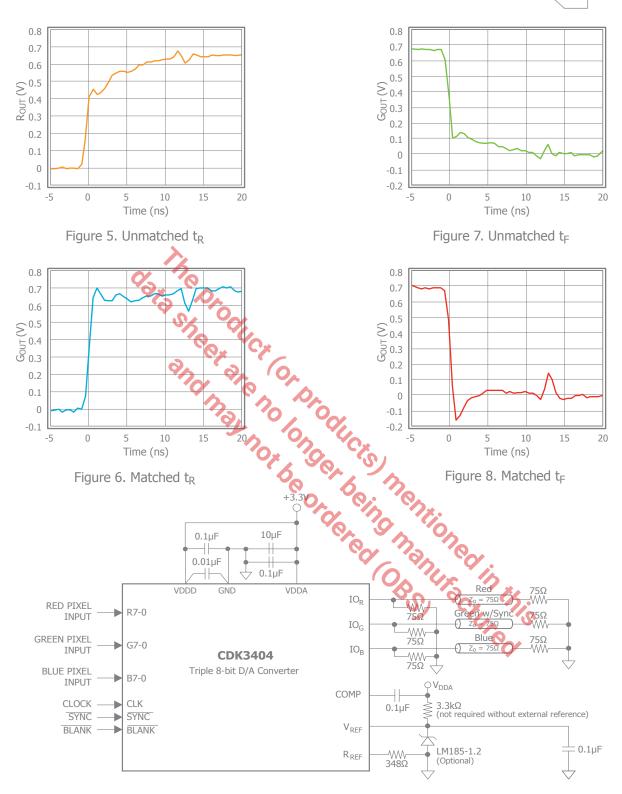


Figure 4. Schematic, Transition Time Sharpening Circuit

A 220nH inductor trims the performance of a 4ft cable, quite well. In Figures xx through xx, the glitch at 12.5ns, is due to a reflection from the source. Not shown, are smaller glitches at 25 and 37.5ns, corresponding to secondary and tertiary reflections. Inductor values should be selected to match the length and type of the cable.





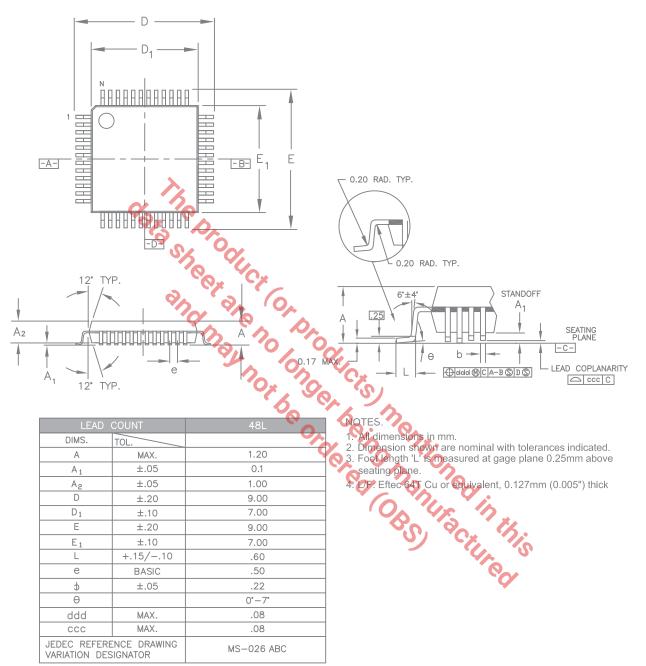
Evaluation boards are available (CEB3400), contact CADEKA for more information.

Related Products

- CDK3400/3401 Triple 10-bit 180MSPS DACs
- CDK3405 Triple 8-bit 180MSPS DAC

Mechanical Dimensions

TQFP-48 Package



For Further Assistance:

Exar Corporation Headquarters and Sales Offices Tel.: +1 (510) 668-7000

Fax: +1 (510) 668-7001 www.exar.com

48720 Kato Road Fremont, CA 94538 - USA



NOTICE

EXAR Corporation reserves the right to make changes to the products contained in this publication in order to improve design, performance or reliability. EXAR Corporation assumes no responsibility for the use of any circuits described herein, conveys no license under any patent or other right, and makes no representation that the circuits are free of patent infringement. Charts and schedules contained here in are only for illustration purposes and may vary depending upon a user's specific application. While the information in this publication has been carefully checked; no responsibility, however, is assumed for inaccuracies.

EXAR Corporation does not recommend the use of any of its products in life support applications where the failure or malfunction of the product can reasonably be expected to cause failure of the life support system or to significantly affect its safety or effectiveness. Products are not authorized for use in such applications unless EXAR Corporation receives, in writing, assurances to its satisfaction that: (a) the risk of injury or damage has been minimized; (b) the user assumes all such risks; (c) potential liability of EXAR Corporation is adequately protected under the circumstances. Reproduction, in part or whole, without the prior written consent of EXAR Corporation is prohibited.

11/11