SP6683

1X/1.5X High Efficiency Charge Pump
LED Driver for Flash or Backlight

FEATURES
- Output Current up to 200mA
- Ideal for White LED driver in CellPhone Flash
- 0.5μA Shutdown Current
- 1x/1.5x Mode Charge Pump
- +2.7V to +5.5V Input Voltage Range
- Built-in 1.2MHz Oscillator
- 0.5mA Quiescent Current
- Over Temperature Protection
- Shorted LED Protection
- Programmable Output Current or Voltage
- PWM Dimming Control via Enable Pin
- Soft Start to limit In-Rush Current
- Space Saving 10-pin 3 x 3mm DFN Package
- Three Options for Optimizing Efficiency:
  - SP6683 - MOSFET discharge in shutdown
  - SP6683-03 - High Impedance in shutdown
  - SP6683-LV - Low Voltage (150mV)VF Reference

DESCRIPTION
The SP6683 is a high power current regulated charge pump ideal for converting a Li-Ion battery input for driving up to 8 white LED’s used in backlighting color displays. The SP6683 operates with an internal 1.2MHz clock, enabling the use of small external components. Output current can be accurately regulated by modulating the switcher between the charge pump and output capacitor. In shutdown mode, the SP6683 discharges the output to ground and draws less than 500nA current. Offered in 10-pin DFN package, the SP6683 family includes three options that allow the user to optimize LED driving circuit; SP6683-LV with low voltage reference, SP6683-03 with high impedance output in shutdown, and SP6683 with pulled down output in shutdown.

APPLICATIONS
- Mobile Phone
- Camera Flash Driver
- PDA
- Digital Still Camera
- Digital Camcorder
- Color LCD Module

TYPICAL APPLICATION SCHEMATIC
### ELECTRICAL CHARACTERISTICS

Unless otherwise specified: \( V_{\text{IN}} = 3.6 \text{V} \), \( C_1 = C_4 = 2.2 \mu\text{F} \) & \( C_2 = C_3 = 1 \mu\text{F} \) (Ceramic, ESR=0.03\( \Omega \)) and \( T_{\text{AMB}} = 25^\circ \text{C} \).

The ♦ denotes specifications which apply over full operating temperature range -40ºC to +85ºC.

#### ABSOLUTE MAXIMUM RATINGS

These are stress ratings only and functional operation of the device at these ratings or any other above those indicated in the operation sections of the specifications below is not implied. Exposure to absolute maximum rating conditions for extended periods of time may affect reliability.

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>Input Voltage</td>
<td>2.7</td>
<td>5.5</td>
<td>V</td>
<td></td>
<td></td>
</tr>
<tr>
<td>Quiescent Current</td>
<td>0.5</td>
<td>3</td>
<td>mA</td>
<td>♦</td>
<td>( V_{\text{IN}} = 3.0 - 5.5\text{V}, V_{\text{OUT}} = 3.6\text{V}, I_{\text{OUT}} = 100\mu\text{A} )</td>
</tr>
<tr>
<td>Shutdown Current</td>
<td>0.5</td>
<td>1.5</td>
<td>( \mu\text{A} )</td>
<td>♦</td>
<td>( V_{\text{EN/PWM}} = 0.0\text{V}, V_{\text{IN}} = 5.5\text{V} )</td>
</tr>
<tr>
<td>Maximum Load Current</td>
<td>200</td>
<td></td>
<td>mA</td>
<td></td>
<td></td>
</tr>
<tr>
<td>Oscillator Frequency</td>
<td>0.7</td>
<td>1.2</td>
<td>1.5 MHz</td>
<td>♦</td>
<td>( V_{\text{IN}} = 2.7 - 5.5\text{V} )</td>
</tr>
<tr>
<td>( V_{\text{FB}} ) Reference Voltage</td>
<td>0.283</td>
<td>0.315</td>
<td>0.347 V</td>
<td>♦</td>
<td>( I_{\text{OUT}} = 20\text{mA}, V_{\text{OUT}} = 3.6\text{V}, 1.5\text{X Mode} )</td>
</tr>
<tr>
<td>Output Resistance</td>
<td>8</td>
<td>13</td>
<td>( \Omega )</td>
<td>♦</td>
<td>1.5X Mode, 100mA Load</td>
</tr>
<tr>
<td>( V_{\text{MODE}} ) Threshold Voltage</td>
<td>1.00</td>
<td>1.25</td>
<td>1.50 V</td>
<td>♦</td>
<td>( V_{\text{IN}} ) Falling</td>
</tr>
<tr>
<td>( V_{\text{MODE}} ) Hysteresis</td>
<td>60</td>
<td>150</td>
<td>mV</td>
<td>♦</td>
<td>( V_{\text{IN}} = 3.6\text{V} )</td>
</tr>
<tr>
<td>( V_{\text{MODE}} ) Pin Current</td>
<td>0.01</td>
<td>0.5</td>
<td>( \mu\text{A} )</td>
<td>♦</td>
<td>( V_{\text{MODE}} = 1.25 \text{V} )</td>
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<tr>
<td>EN/PWM Logic Low</td>
<td>0.4</td>
<td></td>
<td>V</td>
<td></td>
<td></td>
</tr>
<tr>
<td>EN/PWM Logic High</td>
<td>1.6</td>
<td></td>
<td>V</td>
<td></td>
<td></td>
</tr>
<tr>
<td>EN/PWM Pin Current</td>
<td>0.01</td>
<td>0.5</td>
<td>( \mu\text{A} )</td>
<td>♦</td>
<td>( V_{\text{EN/PWM}} = 4.2\text{V} )</td>
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<tr>
<td>FB Pin Current</td>
<td>0.01</td>
<td>0.5</td>
<td>( \mu\text{A} )</td>
<td>♦</td>
<td>( V_{\text{FB}} = 1\text{V} )</td>
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<tr>
<td>( V_{\text{OUT}} ) Ripple</td>
<td>80</td>
<td></td>
<td>mV</td>
<td>♦</td>
<td>( V_{\text{OUT}} = 4\text{V}, I_{\text{OUT}} = 100\text{mA} , 1.5\text{x Mode} )</td>
</tr>
<tr>
<td>( V_{\text{OUT}} ) Turn-On Time</td>
<td>250</td>
<td>500</td>
<td>( \mu\text{s} )</td>
<td>♦</td>
<td>FB within 90% regulation, 1.5X Mode, ( V_{\text{OUT}} = 3.6\text{,} I_{\text{OUT}} = 100\mu\text{A} )</td>
</tr>
</tbody>
</table>
### PIN DESCRIPTION

<table>
<thead>
<tr>
<th>PIN NUMBER</th>
<th>PIN NAME</th>
<th>DESCRIPTION</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>( V_{OUT} )</td>
<td>Regulated charge pump output.</td>
</tr>
<tr>
<td>2</td>
<td>C1P</td>
<td>Positive terminal to the charge pump flying capacitor C2.</td>
</tr>
<tr>
<td>3</td>
<td>( V_{IN} )</td>
<td>Input supply voltage.</td>
</tr>
<tr>
<td>4</td>
<td>( V_{MODE} )</td>
<td>Charge pump mode program pin. When ( V_{MODE} ) is greater than 1.25V, a X1 charge pump is used. Otherwise, charge pump switches to X1.5 mode. A voltage divider shown in the typical application circuit programs the ( V_{IN} ) threshold for charge pump mode switching.</td>
</tr>
<tr>
<td>5</td>
<td>FB</td>
<td>This is the feedback pin for output current or voltage regulation. The voltage of this pin is compared with an internal 315mV reference.</td>
</tr>
<tr>
<td>6</td>
<td>EN/PWM</td>
<td>Enable and PWM dimming control input. Pull this pin low to disconnect ( V_{OUT} ) from ( V_{IN} ) and shutdown the SP6683. ( V_{OUT} ) is pulled to ground in shutdown.</td>
</tr>
<tr>
<td>7</td>
<td>C2N</td>
<td>Negative terminal to the charge pump flying capacitor, C4.</td>
</tr>
<tr>
<td>8</td>
<td>GND</td>
<td>Ground reference.</td>
</tr>
<tr>
<td>9</td>
<td>C1N</td>
<td>Negative terminal to the charge pump flying capacitor, C2.</td>
</tr>
<tr>
<td>10</td>
<td>C2P</td>
<td>Positive terminal to the charge pump flying capacitor C4.</td>
</tr>
</tbody>
</table>

### FUNCTIONAL DIAGRAM

![Functional Diagram](image)

\( V_{IN} \) - Voltage Reference

\( 1.25V \) - Mode Control

\( 1.2 \text{ MHz Clock Manager} \)

\( V_{OUT} \) - Start-up and Charge Pump Switches

\( 315mV \) - Feedback

\( \text{GND} \) - Ground Reference
TYPICAL PERFORMANCE CHARACTERISTICS (25°C)

I_{out} vs. V_{in} @ 200mA

- \(V_f = 3.0V\)
- \(V_f = 3.3V\)
- \(V_f = 3.6V\)
- \(V_f = 3.9V\)
- \(V_f = 4.2V\)

I_{out} vs. V_{in} @ 300mA

- \(V_f = 3.0V\)
- \(V_f = 3.3V\)
- \(V_f = 3.6V\)
- \(V_f = 3.9V\)
- \(V_f = 4.2V\)

I_{out} vs. V_{in} @ 60mA

- \(V_f = 3.0V\)
- \(V_f = 3.3V\)
- \(V_f = 3.6V\)
- \(V_f = 3.9V\)
- \(V_f = 4.2V\)

I_{out} vs. V_{in} @ 100mA

- \(V_f = 3.0V\)
- \(V_f = 3.3V\)
- \(V_f = 3.6V\)
- \(V_f = 3.9V\)
- \(V_f = 4.2V\)
TYPICAL PERFORMANCE CHARACTERISTICS (25°C)

Efficiency vs Vin (Iout = 60mA)

Efficiency vs Vin (Iout = 100mA)

Efficiency vs Vin (Iout = 200mA)

Efficiency vs Vin (Iout = 300mA)
General Overview

The SP6683 is a current regulated charge pump ideal for converting a Li-Ion battery input for driving white LEDs used in camera flash or backlighting color displays in cellular phones, PDAs, digital cameras and MP3 players. The SP6683 is able to efficiently drive one 200mA Flash LED or up to eight 20mA white LEDs in parallel and maintain a constant brightness over a very wide operating voltage range (2.7V to 5.5V). The SP6683 operates with an internal 1.2MHz clock, enabling the use of small external components. Other features of SP6683 include PWM dimming control as well as complete input/output disconnect in shutdown. In shutdown mode the IC typically draws 500nA current. The output regulation is achieved by sensing the voltage at the feedback pin and modulating the switch between the charge pump and output capacitor.

Theory of Operation

The SP6683 regulated charge pump block diagram consists of four main blocks (Voltage Reference, Mode Control, Clock Manager, Startup and Charge-Pump Switches) and two comparators (V_{MODE} Comparator and V_{OUT} Comparator).

1) Voltage Reference. This block provides the 315mV and 1.25V reference voltages needed for the two comparators.

2) Mode Control. An external voltage divider connected to the V_{MODE} pin will define an input voltage to the mode comparator which sets the logic state of the mode selection outputs to the X1 or X1.5 modes. V_{MODE} is compared to a 1.25V bandgap voltage. For example, if one makes a 158K/100K divider, the mode will change at 2.58 x 1.25 V = 3.23V. A comparator-based cycle by cycle regulation ensures that no mode change occurs during cycles.

3) Clock Manager. An internal 1.2MHz clock is generated in this block. Depending on the mode control, the appropriate clock phasing is generated here and sent to the startup and charge pump switches block.

4) V_{OUT} Comparator & Output Control. A 315mV reference voltage is compared to feedback voltage to control the V_{OUT} needed for the application. Output current is set by a bias resistor from the FB pin to the GND pin chosen by the relationship:

\[ I_{OUT} = \frac{V_{FB}}{R_{FB}} \]

where \( V_{FB} = 315 \text{mV} \).
Configuring as Cellphone Camera Flash Driver

SP6683 is capable of delivering up to 200mA current, which is perfect to drive the most common LEDs in today's market. It is designed to work in the cell phone camera flash which powers up directly from the Lithium-Ion battery.

![Camera Flash Application](image)

Configuring as LCD Backlight

The SP6683 is able to efficiently drive up to eight 20mA white LEDs in parallel and maintain a constant brightness over a very wide operating voltage range (2.7V to 5.5V).

![LCD Backlight Application](image)

Configuring as Camera Flash + Torch Driver

The SP6683 can be configured as a flash/torch LED driver which in some applications require constant torch current used for motion pictures (i.e. movie mode).

![Camera Flash/Torch Application](image)

Programming X1 or X1.5 Operating Mode

For simplest application ciruitry, SP6683’s mode can be fixed by connecting V\text{MODE} to V\text{IN} for X1 mode or connecting V\text{MODE} to GND for X1.5 mode. The SP6683 can automatically change from X1 mode to X1.5 to maximize efficiency. To use this feature, divider resistors should be chosen according to the specific application, as shown below.

![Programming the Vmode Resistors](image)
Configuring the SP6683 as Voltage or Current Source

The white LED load configuration used by customers can be discrete white LEDs or a white LED module. Inside the white LED module, there may or may not be resistors in series with the white LEDs. According to the different application requirements, the SP6683 can be configured as either a voltage source or a current source to provide solutions for these different applications, as shown below. The figure shows the SP6683 driving discrete white LEDs as a current source.

\[
R_B = \frac{V_{FB}}{I_{LED}}
\]

The current in one white LED current is set by the ratio of the feedback pin voltage (315mV) and the bias resistor \( R_B \). To set the operating current, \( R_B \) can be selected by:

Driving discrete white LEDs as current source

Driving 2-wire white LED module as current source

In the figure above, the SP6683 was used to drive a 2-wire white LED module (without internal series resistors) as a current source. The bias resistor \( R_B \) is selected to regulate the total current of the white LED module instead of the current of a single LED as in the figure showing the 3 wire white LED module.

Driving 2-wire white LED module as voltage source

Driving 3-wire white LED module as voltage source
In this application, the bias resistor can be selected by:

\[ R_B = \frac{V_{FB}}{I_{LED(TOTAL)}} \]

where \( I_{LED(TOTAL)} \) is the total operating current of all the white LEDs.

To use SP6683 as a voltage source for fixed voltage applications, a voltage divider is needed to program the output voltage, as shown in the figure above. The output voltage is set by the ratio of the two resistors and the feedback control voltage as shown by:

\[ V_{OUT} = (1 + \frac{R_5}{R_6}) \cdot V_{FB} \]

**Brightness Control Using PWM**

Dimming control can be achieved by applying a PWM control signal to the EN/PWM pin. The brightness of the white LEDs is controlled by increasing and decreasing the duty cycle of the PWM signal. While operating frequency range is from 60Hz to 700Hz, the recommended maximum brightness frequency range is from 60Hz to 200Hz. A repetition rate of at least 60Hz is required to prevent flicker.

**Three Options for Optimization**

The SP6683 family includes three options that allow the user to optimize the LED driving circuit. The SP6683 includes an on-board discharge switch that ensures that \( V_{OUT} \) is completely discharged in shutdown. This feature is very useful in systems using PWM dimming, when the SP6683 is periodically turned on and off. The discharge switch ensures that the falling and rising edges of the LED current have similar durations.

The SP6683-03ER has the discharge MOSFET disabled ensuring that the output node is at high impedance during shutdown. This can be useful if the system is feeding LED current from another source while the SP6683 is in shutdown.

The SP6683-LV offers a lower reference voltage that allows to reduce efficiency loss occurring in the current sense resistor. This part is ideal for driving single LEDs for Flash applications.

**Reducing Component Count**

It is possible to reduce the number of external components by shorting pins 2 & 9 and omitting the flying capacitor. In this mode, the charge pump operates in a "quasi" doubler mode. This reduces efficiency but does allow for one less external capacitor. This circuit is recommended only when Input and Output Voltages have similar range, which is very common when driving white LEDs from a Li-Io battery.

**Flying Cap Removed**

In this application, the bias resistor can be selected by:

\[ R_B = \frac{V_{FB}}{I_{LED(TOTAL)}} \]

where \( I_{LED(TOTAL)} \) is the total operating current of all the white LEDs.

To use SP6683 as a voltage source for fixed voltage applications, a voltage divider is need to program the output voltage, as shown in the figure above. The output voltage is set by the ratio of the two resistors and the feedback control voltage as shown by:

\[ V_{OUT} = (1 + \frac{R_5}{R_6}) \cdot V_{FB} \]
Component Selection

The guideline for selecting divider resistors is as follows. For high input voltage, the SP6683 will work in X1 mode. When the input voltage drops to \( V_{TH} \) threshold voltage, it will switch to X1.5 mode automatically. The \( V_{TH} \) threshold voltage for mode change can be calculated by:

\[
V_{TH} = (V_F + 0.315 + m \cdot I_{LED} \cdot R_{OUT})
\]

Where \( V_F \) and \( m \) are the forward voltage and number of the white LEDs, \( R_{OUT} \) is the output resistance of the SP6683.

The equation for the voltage divider \( R_1 \) and \( R_2 \) with \( V_{MODE} = 1.25V \) is:

\[
V_{TH} = 1.25V \cdot (1+R_1/R_2)
\]

Which can be expressed as \( R_1 \):

\[
R_1 = \left[ \frac{V_{TH}}{1.25} - 1 \right] R_2
\]

For the typical SP6683 application, using \( V_F = 3.6V, m = 8, I_{LED} = 15mA, R_{OUT} = 6\Omega, V_{TH} \) will be 4.63V, Select \( R_2 = 100k\Omega \), then \( R_1 = 270k\Omega \)

Ceramic capacitors are recommended for their inherently low ESR, which will help produce low peak-to-peak output ripple, and reduce high frequency spikes.

The fly capacitor controls the strength of the charge pump. Selection of the fly capacitor is a tradeoff between the output voltage ripple and the output current capability. Decreasing the fly capacitor will reduce the output voltage ripple because less charge will be delivered to the output capacitor. However, a smaller fly capacitor leads to larger output resistance, thus decreasing the output current capability and the efficiency of the circuit. Increasing the value of the input and output capacitors could further reduce the input and output ripple. For backlighting applications up to 80mA output current, \( C_{IN} = C_{OUT} = 1\mu F \) is typically used. For applications with more than 80mA output current, for example in Flash applications where \( I_{out} \) can be as much as 200mA, \( C_{IN} = C_{OUT} = 2.2\mu F \) is recommended. For peak performance, place all capacitors as close to the SP6683 as possible.

Refer to table 1 below for some suggested low ESR capacitors.

<table>
<thead>
<tr>
<th>Manufacturers/Website</th>
<th>Part Number</th>
<th>Capacitance/ Voltage</th>
<th>Capacitor Size/Type/Thickness</th>
<th>ESR at 100KHz</th>
</tr>
</thead>
<tbody>
<tr>
<td>TDK/www.tdk.com</td>
<td>C1005X5R0J105M</td>
<td>1uF/6.3V</td>
<td>0402/X5R/0.5mm</td>
<td>0.03</td>
</tr>
<tr>
<td>TDK/www.tdk.com</td>
<td>C1608X5R0J225K</td>
<td>2.2uF/6.3V</td>
<td>0603/X5R/0.9mm</td>
<td>0.02</td>
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<tr>
<td>Murata/www.murata.com</td>
<td>GRM155R60J105KE19B</td>
<td>1uF/6.3V</td>
<td>0402/X5R/0.55mm</td>
<td>0.03</td>
</tr>
<tr>
<td>Murata/www.murata.com</td>
<td>GRM188R61A225KE34B</td>
<td>2.2uF/6.3V</td>
<td>0603/X5R/0.9mm</td>
<td>0.02</td>
</tr>
</tbody>
</table>
**Brightness Matching**

For white LEDs, the forward voltage drop is a function of the operating current. However, for a given current, the forward voltage drop across each LED does not always match exactly. This can cause uneven brightness in the white LEDs.

In the figure below, assuming high-precision bias resistors were used, the operating current ratio of two different branches can be easily derived as shown by:

\[
\frac{I_1}{I_2} = \frac{V_{OUT} - V_{F1}}{V_{OUT} - V_{F2}}
\]

where \(I_1\) & \(I_2\) are the operating current of the white LEDs, where \(V_{F1}\) & \(V_{F2}\) are the forward voltage of the white LEDs.

Since the brightness of the white LED is proportional to the operating current, for better brightness matching, a higher output voltage could be used. This could be done by using a larger resistor, as shown in the figure below. \(R_{b2}\) is used to bias the operating current of the white LED, \(R_{b1}\) is used to increase the output voltage. Better brightness matching was achieved at the cost of the power wasted on the bias resistor.

**Power Efficiency**

The efficiency of driving the white LEDs can be calculated by

\[
\eta = \frac{V_F \cdot I_F}{V_i \cdot I_i} = \frac{V_F \cdot I_F}{V_i \cdot (n \cdot I_F + I_Q)} \approx \frac{V_F}{V_i \cdot n}
\]

Where

\(V_i\) & \(I_i\) are input voltage and current;
\(V_F\) & \(I_F\) are the forward voltage and operating current of the White LED;
\(I_Q\) is quiescent current, which is considered small compared with \(I_F\).
SP6683 Li-Ion Input in 1x to 1.5X mode to LumiLED PWF1

Vin: 3.0 - 4.2V

R1 = (Vthres/1.25 - 1)*R2
R1 = 174K for 100mA Vthres = 3.4V
R1 = 205K for 200mA Vthres = 3.8V

Efficiency: 1X-1.5X Cin = Cout = 2.2uF

Output Current: 1X-1.5X Cin = Cout = 2.2uF
### PACKAGE: 10 PIN DFN

**SIDE VIEW**

**TOP VIEW**

**BOTTOM VIEW**

<table>
<thead>
<tr>
<th>SYMBOL</th>
<th>Dimensions in Millimeters: Controlling Dimension</th>
<th>Dimensions in Inches</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>MIN</td>
<td>NOM</td>
</tr>
<tr>
<td>A</td>
<td>0.80</td>
<td>0.90</td>
</tr>
<tr>
<td>A1</td>
<td>0.00</td>
<td>0.02</td>
</tr>
<tr>
<td>A3</td>
<td>0.20 REF</td>
<td></td>
</tr>
<tr>
<td>K</td>
<td>0.20</td>
<td></td>
</tr>
<tr>
<td>ø</td>
<td>0º</td>
<td></td>
</tr>
<tr>
<td>b</td>
<td>0.18</td>
<td>0.25</td>
</tr>
<tr>
<td>D</td>
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</tr>
<tr>
<td>D2</td>
<td>2.20</td>
<td></td>
</tr>
<tr>
<td>E</td>
<td>3.00 BSC</td>
<td></td>
</tr>
<tr>
<td>E2</td>
<td>1.40</td>
<td></td>
</tr>
<tr>
<td>e</td>
<td>0.50 BSC</td>
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</tr>
<tr>
<td>L</td>
<td>0.30</td>
<td>0.40</td>
</tr>
</tbody>
</table>

SIPEX Pkg Signoff Date/Rev: JL Aug09-05 / RevA

Date: 3/31/06 RevG

SP6683 1X/1.5X Charge Pump LED Driver for Flash or Backlight

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<table>
<thead>
<tr>
<th>Part Number</th>
<th>Operating Temperature Range</th>
<th>Package Type</th>
</tr>
</thead>
<tbody>
<tr>
<td>SP6683ER</td>
<td>-40°C to +85°C</td>
<td>10 Pin DFN</td>
</tr>
<tr>
<td>SP6683ER/TR</td>
<td>-40°C to +85°C</td>
<td>10 Pin DFN</td>
</tr>
<tr>
<td>SP6683-03ER</td>
<td>-40°C to +85°C</td>
<td>10 Pin DFN</td>
</tr>
<tr>
<td>SP6683-03ER/TR</td>
<td>-40°C to +85°C</td>
<td>10 Pin DFN</td>
</tr>
<tr>
<td>SP6683ER-L</td>
<td>-40°C to +85°C</td>
<td>10 Pin DFN</td>
</tr>
<tr>
<td>SP6683ER-L/TR</td>
<td>-40°C to +85°C</td>
<td>10 Pin DFN</td>
</tr>
<tr>
<td>SP6683-03ER-L</td>
<td>-40°C to +85°C</td>
<td>10 Pin DFN</td>
</tr>
<tr>
<td>SP6683-03ER-L/TR</td>
<td>-40°C to +85°C</td>
<td>10 Pin DFN</td>
</tr>
</tbody>
</table>

*Contact factory for availability of the following parts*

<table>
<thead>
<tr>
<th>Part Number</th>
<th>Operating Temperature Range</th>
<th>Package Type</th>
</tr>
</thead>
<tbody>
<tr>
<td>SP6683LVER</td>
<td>-40°C to +85°C</td>
<td>10 Pin DFN</td>
</tr>
<tr>
<td>SP6683LVER/TR</td>
<td>-40°C to +85°C</td>
<td>10 Pin DFN</td>
</tr>
</tbody>
</table>

/TR = Tape and Reel

Pack quantity is 3000 for DFN.